
Model-Aided Data Driven Adaptive Target Detection for Channel Matrix-Based Cognitive Radar

Touseef Ali, Akshay S. Bondre, and Christ D. Richmond

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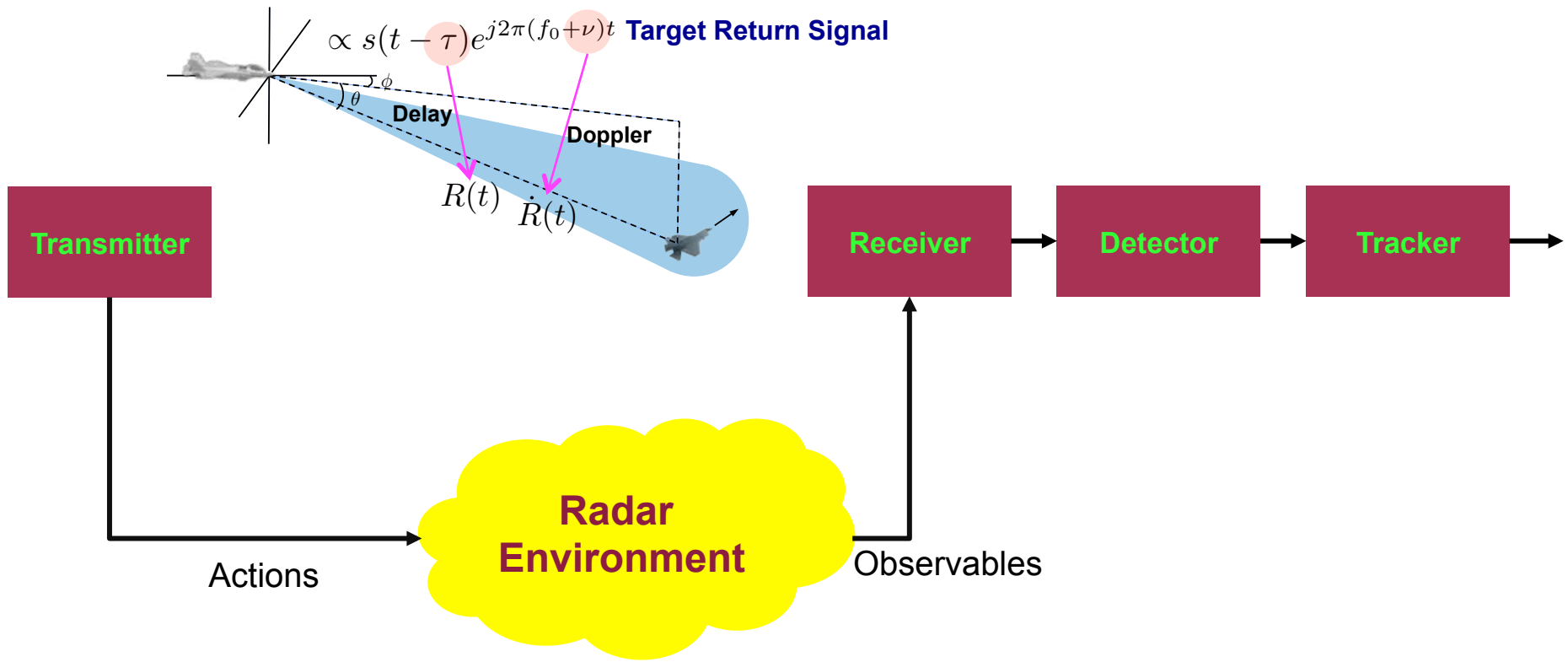
Outline



- • **Adaptive Target Detection for Conventional Radar**
- Adaptive Target Detection for Cognitive Radar
- Deep Neural Network (DNN) based Target Detection
- Summary



Conventional Radar: A Feed-Forward System

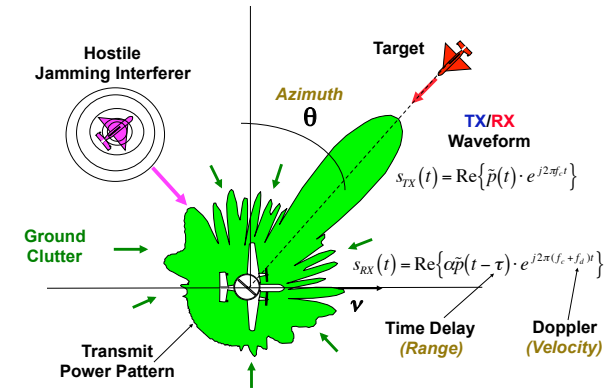




Conventional Radar Data Model



- Radar with array of l elements transmitting waveform $s(t)$
- Radar data often modeled as complex Gaussian: $\mathbf{x} \sim \mathcal{CN}(\alpha \cdot \mathbf{v}, \mathbf{R})$
- $E\{\mathbf{x}\} = \alpha \cdot \mathbf{v} \rightarrow$ **Signal (Target)**
 - $\alpha = \alpha[s(t)]$ **Target complex amplitude**
 - \mathbf{v} **Target space-time response:** $\mathbf{v}_{RADAR} = \mathbf{v}(\varphi, \theta, f_d)$,
 - Azimuth, Elevation
 - Doppler
- $\text{cov}(\mathbf{x}) = \mathbf{R} \rightarrow$ **Noise + Interference**
 - $\mathbf{R} = \mathbf{R}_{System} + \mathbf{R}_{Jammer} + \mathbf{R}_{Clutter}$
 - $\mathbf{R}_{Clutter} = \mathbf{R}_{Clutter}[s(t)] \rightarrow$ **Depends on Tx Waveform**
 - **Thus, waveform dependence is nonlinear**





Conventional Model: Adaptive Array Detection Problem



- Radar with array of l elements
- **Test** data vector \mathbf{x}
- Look for presence of target in \mathbf{x} :

$$H_0 : \mathbf{x} = \mathbf{n}, \quad \mathbf{x} \sim \mathcal{CN}(\mathbf{0}, \mathbf{R})$$

$$H_1 : \mathbf{x} = \alpha \cdot \mathbf{v} + \mathbf{n}, \quad \mathbf{x} \sim \mathcal{CN}(\alpha \cdot \mathbf{v}, \mathbf{R})$$
- Unknowns are α and \mathbf{R} with possible error in \mathbf{v}

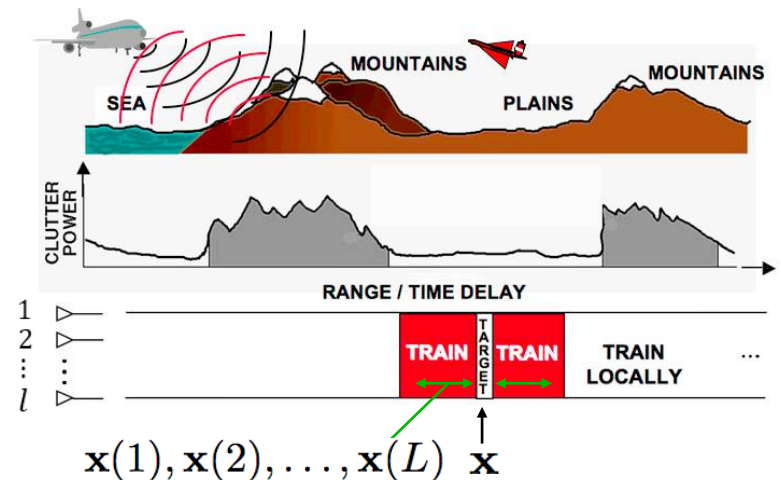
- L **Training** noise only data vectors

$$\mathbf{x}(1), \mathbf{x}(2), \dots, \mathbf{x}(L) \quad L \geq l$$

$$\mathbf{x}(k) \sim \mathcal{CN}(\mathbf{0}, \mathbf{R})$$

- **Desire viable detection statistic:** $t(\mathbf{x}, \mathbf{x}(1), \mathbf{x}(2), \dots, \mathbf{x}(L), \mathbf{v})$

Example: Radar Environment





Conventional Model: Summary of Adaptive Detection Algorithms



- Adaptive Matched Filter (AMF)

Robey, et. al. IEEE T-AES 1992

Reed & Chen 1992, Reed et. al. 1974



$$t_{AMF} = \frac{|\mathbf{v}^H \hat{\mathbf{R}}^{-1} \mathbf{x}|^2}{\mathbf{v}^H \hat{\mathbf{R}}^{-1} \mathbf{v}}$$

- Generalized Likelihood Ratio Test (GLRT)

Kelly IEEE T-AES 1986, Khatri & Rao 1985



$$t_{GLRT} = \frac{t_{AMF}}{1 + \mathbf{x}^H \hat{\mathbf{R}}^{-1} \mathbf{x}}$$

- Adaptive Coherence Estimator (ACE)

Conte et. al. IEEE T-AES 1995,

Scharf Asil. 1996, Kraut IEEE T-SP 2001



$$t_{ACE} = \frac{t_{AMF}}{\mathbf{x}^H \hat{\mathbf{R}}^{-1} \mathbf{x}}$$

- Adaptive Sidelobe Blanker (ASB)

Kreithen, Baranoski, 1996

Richmond Asilomar 1997& 1998,

Richmond IEEE T-SP 2000



$$f(t_{AMF}, t_{ACE})$$

Each Algorithm is Function of
Sample Covariance



$$\hat{\mathbf{R}} = \mathbf{x}(1)\mathbf{x}^H(1) + \dots + \mathbf{x}(L)\mathbf{x}^H(L)$$



Outline



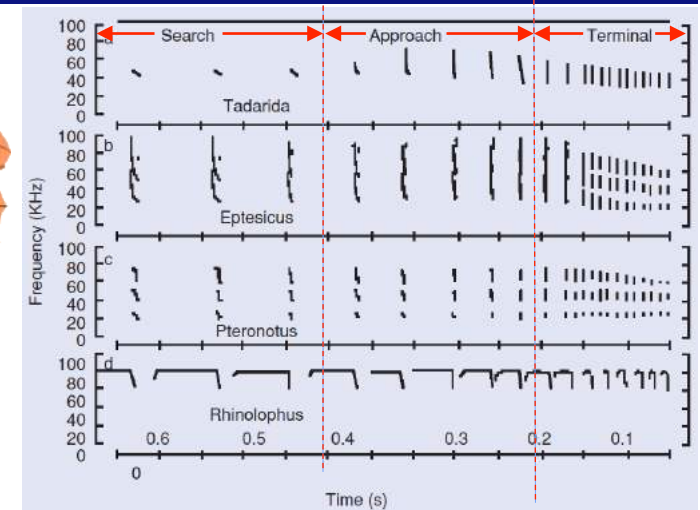
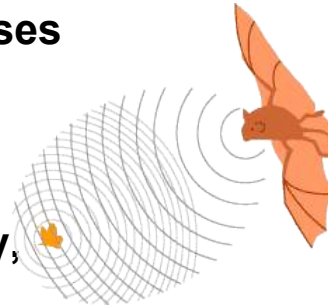
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- • **Adaptive Target Detection for Cognitive Radar**
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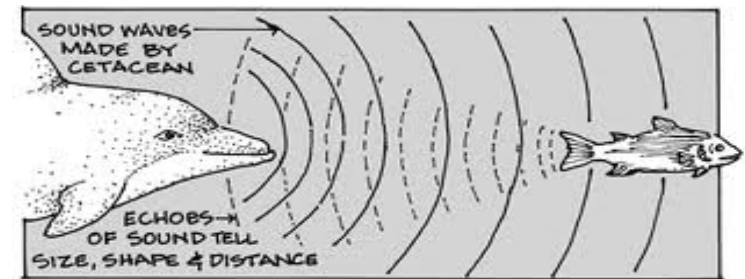
Inspiration for Cognitive Systems: Echo Location / Human Vision-Perception



- **Bats and dolphins** emit short sound pulses to locate food
- Return echoes inform about type of prey, range and bearing
- **Waveform characteristics adapted** as bat / dolphin closes on prey
 - Higher pulse repetitions
 - Changes in chirp rates
- **Human visual brain**
 - Cerebral cortex major player in cognition (Fuster model introduced)



S. Haykin, "Cognitive radar: a way of the future,"
IEEE Sig. Proc. Mag., January 2006.

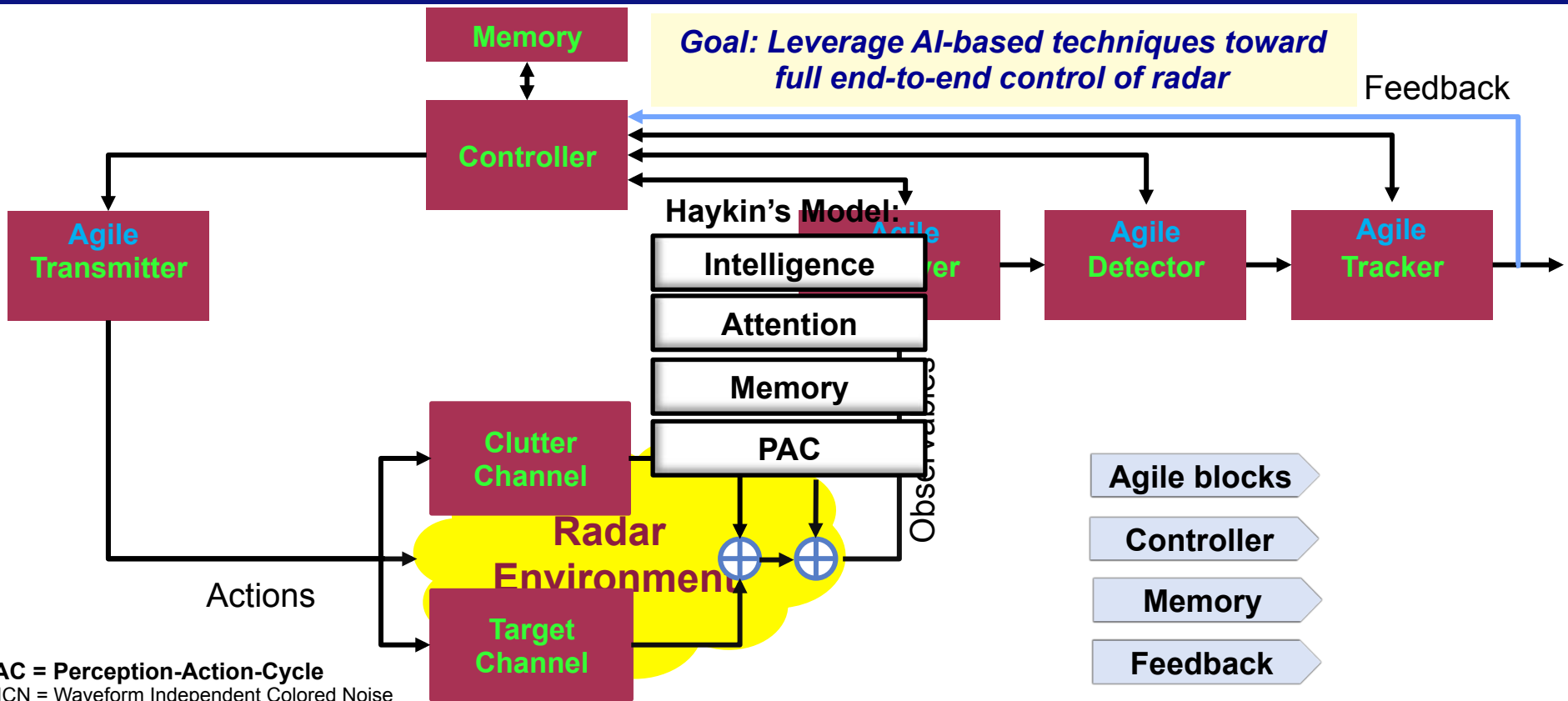




Cognitive Radar: Adaptation via a Feedback System



Goal: Leverage AI-based techniques toward full end-to-end control of radar



PAC = Perception-Action-Cycle
WICN = Waveform Independent Colored Noise



Motivating Channel Matrix-Based Radar Model



- **LTV system output:** $r(t) = \int \tilde{h}(t, a) s(a) da \stackrel{\Delta}{=} - \int \tilde{h}(t, t - \tau) s(t - \tau) d\tau \stackrel{\Delta}{=} \int h(t, \tau) s(t - \tau) d\tau$
 $\underbrace{a = t - \tau, da = -d\tau}_{\uparrow}$, $\underbrace{h(t, \tau) \stackrel{\Delta}{=} -\tilde{h}(t, t - \tau)}_{\uparrow}$
- **Consider Fourier representation** $h(t, \tau) = \int H(\tau, \nu) e^{j2\pi\nu t} d\nu \implies r(t) = \int \int H(\tau, \nu) s(t - \tau) e^{j2\pi\nu t} d\nu d\tau$
- **Thus,** $r(t) \simeq \sum_i \sum_m H_{i,m} s(t - \tau_i) e^{j2\pi\nu_m t}$. **Discretizing yields form*** $r[n] = \sum_i \sum_m \mathcal{H}[i, m] s[n - i] e^{j\frac{2\pi m}{M} n}$
- **Note that** $r[n] = \sum_k \sum_m \mathcal{H}[n - k, m] s[k] e^{j\frac{2\pi m}{M} n} = \sum_k \left(\sum_m \mathcal{H}[n - k, m] e^{j\frac{2\pi m}{M} n} \right) \cdot s[k] \stackrel{\Delta}{=} \sum_k \tilde{\mathcal{H}}_n[k] s[k]$
 $\underbrace{k = n - i, i = n - k}_{\uparrow}$
- **Each received data sample is linear transformation of waveform:**

$$\begin{bmatrix} r[0] \\ r[1] \\ \vdots \\ r[N-1] \end{bmatrix} = \underbrace{\begin{bmatrix} * & \cdots & * \\ \vdots & \ddots & \vdots \\ * & \cdots & * \end{bmatrix}}_{\text{Channel Matrix}} \underbrace{\begin{bmatrix} s[0] \\ s[1] \\ \vdots \\ s[N-1] \end{bmatrix}}_{\text{Waveform}} + \underbrace{\begin{bmatrix} \tilde{n}[0] \\ \tilde{n}[1] \\ \vdots \\ \tilde{n}[N-1] \end{bmatrix}}_{\text{Noise + Interference}} \implies \boxed{\mathbf{r} = \mathbf{H}\mathbf{s} + \mathbf{n}}$$

- **Linear model can capture dominant effects of doubly spread multipath channel**



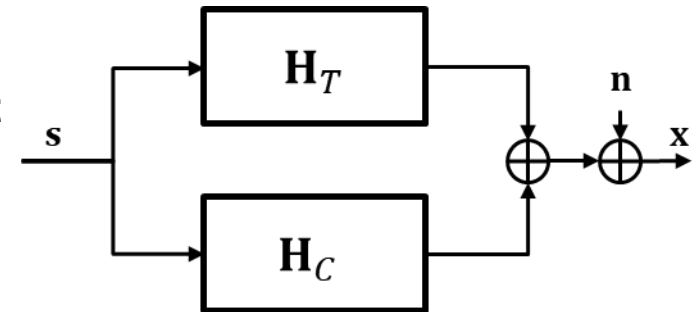
Channel Matrix-based Data Model Used in Cognitive Radar*



- **Target** produces scattering component, say $H_T s$
- **Clutter / reverberation** also yields scattering component
 - Can model in similar way, i.e. as $H_C s$
- Thus, we have the **general MIMO radar data model**:

$$\mathbf{x} = \mathbf{H}_T \mathbf{s} + \mathbf{H}_C \mathbf{s} + \mathbf{n}$$

- **Same TX waveforms** produce both target return and clutter
- **Scattering propagation paths are different, however, i.e. $H_T \neq H_C$**
- **Simplifies waveform optimization problem**
- **Although model is linear, fundamental research questions remained unaddressed**





Cognitive Radar Target Detection Problem



- **Binary hypothesis test:**
 - No Target $\implies H_0 : \mathbf{x} = \mathbf{H}_C \mathbf{s} + \mathbf{n}$, $\mathbf{x} \sim \mathcal{CN}(\mathbf{H}_C \mathbf{s}, \mathbf{R})$
 - Target Present $\implies H_1 : \mathbf{x} = \mathbf{H}_C \mathbf{s} + \mathbf{H}_T \mathbf{s} + \mathbf{n}$, $\mathbf{x} \sim \mathcal{CN}(\mathbf{H}_C \mathbf{s} + \mathbf{H}_T \mathbf{s}, \mathbf{R})$
- **Waveform independent colored noise (WICN)** $\mathbf{n} \sim \mathcal{CN}(\mathbf{0}, \mathbf{R})$
- $\mathbf{H}_T, \mathbf{H}_C$ and \mathbf{R} known
 - Likelihood Ratio Test (LRT)
- $\mathbf{H}_T, \mathbf{H}_C$ and \mathbf{R} unknown \rightarrow **Composite Hypothesis testing**
 - Average Likelihood Ratio Test (ALRT)*
 - Channel matrices modeled as random
 - Generalized Likelihood Ratio Test (GLRT)**
 - Channel matrices assumed deterministic but unknown during the observation time

GLRT and ALRT establish benchmarks for comparison with machine / deep learning-based cognitive radar detection, and guides AI architectures



Cognitive Radar Data Model: Adaptive Detection Problem Formulation



- Motivated by GLRT approach taken by Kelly*
- K **secondary** and M **primary data vectors**, $L_s = K + M$.
- **Total received data matrix** $\mathbf{X} \in \mathbb{C}^{l \times L_s}$

$$\mathbf{X} = \underbrace{[\mathbf{x}(1) | \cdots | \mathbf{x}(K)]}_{\substack{\text{Secondary Data} \\ \text{(Clutter + noise only)}}} \underbrace{[\mathbf{x}(K+1) | \cdots | \mathbf{x}(L_s)]}_{\substack{\text{Primary Data} \\ \text{(May contain target as well)}}} = [\mathbf{X}_S | \mathbf{X}_P] \quad \mathbf{X}_S \in \mathbb{C}^{l \times K} \quad \mathbf{X}_P \in \mathbb{C}^{l \times M}$$
- **Secondary data** $\rightarrow \mathbf{x}(k) = \mathbf{H}_C \mathbf{s}(k) + \mathbf{n}(k), k = 1, \dots, K$
- **Primary data** $\rightarrow \left. \begin{array}{l} H_0 : \mathbf{x}(k) = \mathbf{H}_C \mathbf{s}(k) + \mathbf{n}(k) \\ H_1 : \mathbf{x}(k) = \mathbf{H}_C \mathbf{s}(k) + \mathbf{H}_T \mathbf{s}(k) + \mathbf{n}(k) \end{array} \right\} k = K + 1, \dots, L_s$
- **Waveform matrix** $\mathbf{S} = [\mathbf{s}(1) | \cdots | \mathbf{s}(K) | \mathbf{s}(K+1) | \cdots | \mathbf{s}(L_s)] = [\mathbf{S}_S | \mathbf{S}_P] \in \mathbb{C}^{l \times L_s}$
- **WICN** $\mathbf{n}(k)$ I.I.D with $\mathbf{n}(k) \sim \mathcal{CN}(\mathbf{0}, \mathbf{R}), k = 1, \dots, L_s$.



GLRT for Cognitive Radar Data Model: Unknown \mathbf{R} , \mathbf{H}_T & \mathbf{H}_C



- Channel matrices $\mathbf{H}_C, \mathbf{H}_T$ and WICN covariance \mathbf{R} are **all unknown**

- The GLRT is given by: $\Lambda_{GLRT}(\mathbf{X}) = \frac{\max_{\mathbf{R}, \mathbf{H}_C, \mathbf{H}_T} f_1(\mathbf{X}; \mathbf{R}, \mathbf{H}_C, \mathbf{H}_T)}{\max_{\mathbf{R}, \mathbf{H}_C} f_0(\mathbf{X}; \mathbf{R}, \mathbf{H}_C)}$

- PDF under H_0

$$f_0(\mathbf{X}; \mathbf{H}_C, \mathbf{R}) = \left[\frac{1}{\pi^l |\mathbf{R}|} \right]^{L_s} \exp \left\{ - \sum_{k=1}^{L_s} (\mathbf{x}(k) - \mathbf{H}_C \mathbf{s}(k))^H \mathbf{R}^{-1} (\mathbf{x}(k) - \mathbf{H}_C \mathbf{s}(k)) \right\}.$$

- PDF under H_1

$$f_1(\mathbf{X}; \mathbf{H}_C, \mathbf{H}_T, \mathbf{R}) = \left[\frac{1}{\pi^l |\mathbf{R}|} \right]^{L_s} \exp \left\{ - \sum_{k=1}^K (\mathbf{x}(k) - \mathbf{H}_C \mathbf{s}(k))^H \mathbf{R}^{-1} (\mathbf{x}(k) - \mathbf{H}_C \mathbf{s}(k)) \right. \\ \left. - \sum_{k=K+1}^{L_s} (\mathbf{x}(k) - (\mathbf{H}_C + \mathbf{H}_T) \mathbf{s}(k))^H \mathbf{R}^{-1} (\mathbf{x}(k) - (\mathbf{H}_C + \mathbf{H}_T) \mathbf{s}(k)) \right\}$$



GLRT for Cognitive Radar Data Model: Unknown \mathbf{R}, \mathbf{H}_T & \mathbf{H}_C^*



$$\tilde{\Lambda}_{GLRT}(\mathbf{X}) = \frac{|\mathbf{X}\mathfrak{P}(\mathbf{S}^H|\mathbf{I})\mathbf{X}^H|}{|\mathbf{X}_P\mathfrak{P}(\mathbf{S}_P^H|\mathbf{I})\mathbf{X}_P^H + \mathbf{X}_S\mathfrak{P}(\mathbf{S}_S^H|\mathbf{I})\mathbf{X}_S^H|} \begin{matrix} H_1 \\ > \\ < \\ H_0 \end{matrix} \tilde{\eta}_{GLRT}$$

- With noiseless input data $\mathbf{X}_P = (\mathbf{H}_C + \mathbf{H}_T)\mathbf{S}_P$, $\mathbf{X}_S = \mathbf{H}_C\mathbf{S}_S \implies$
 - $\mathbf{X}_S\mathfrak{P}(\mathbf{S}_S^H|\mathbf{I})\mathbf{X}_S^H$ nullifies everything in clutter subspace in sec. data
 - $\mathbf{X}_P\mathfrak{P}(\mathbf{S}_P^H|\mathbf{I})\mathbf{X}_P^H$ nullifies everything in clutter+target subspaces in pri. data
 - $\mathbf{X}\mathfrak{P}(\mathbf{S}^H|\mathbf{I})\mathbf{X}^H$ cancels everything in clutter subspaces in both pri. and sec. data sets and in target subspace in pri. data set except the residue: $\mathbf{H}_T\mathbf{S}_P[\mathbf{I} - \mathbf{S}_P^H(\mathbf{S}\mathbf{S}^H)^{-1}\mathbf{S}_P]\mathbf{S}_P^H\mathbf{H}_T^H$

Denominator approximates $|\mathbf{R}|$

- Maximum-likelihood estimates of clutter channel matrices:

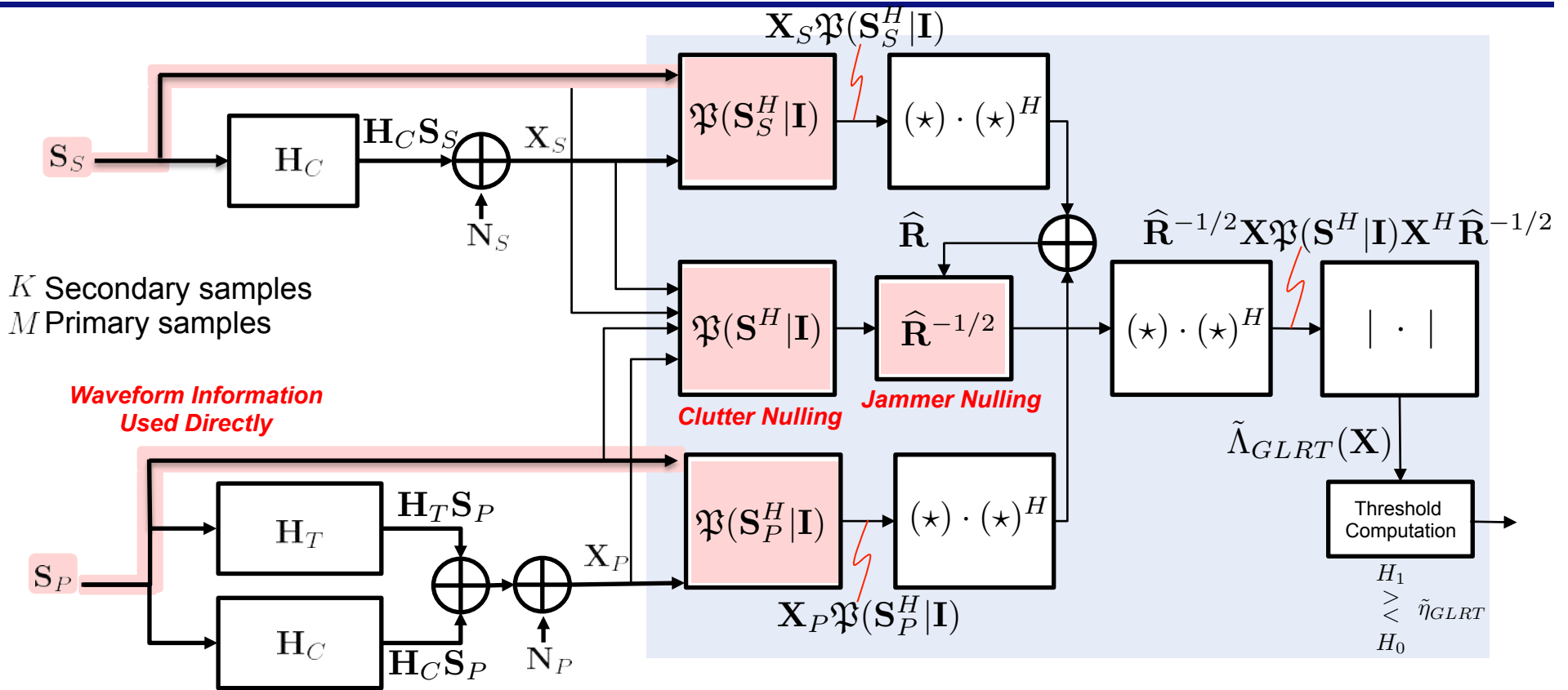
$$\hat{\mathbf{H}}_{C0} = \mathbf{X}\mathbf{S}^H(\mathbf{S}\mathbf{S}^H)^{-1} \quad (\text{under } H_0)$$

$$\hat{\mathbf{H}}_{C1} = \mathbf{X}_S\mathbf{S}_S^H(\mathbf{S}_S\mathbf{S}_S^H)^{-1} \quad (\text{under } H_1)$$

- GLRT statistic depends only on measured data and waveform (desire to optimize)



GLRT Architecture for Cognitive Radar Detection





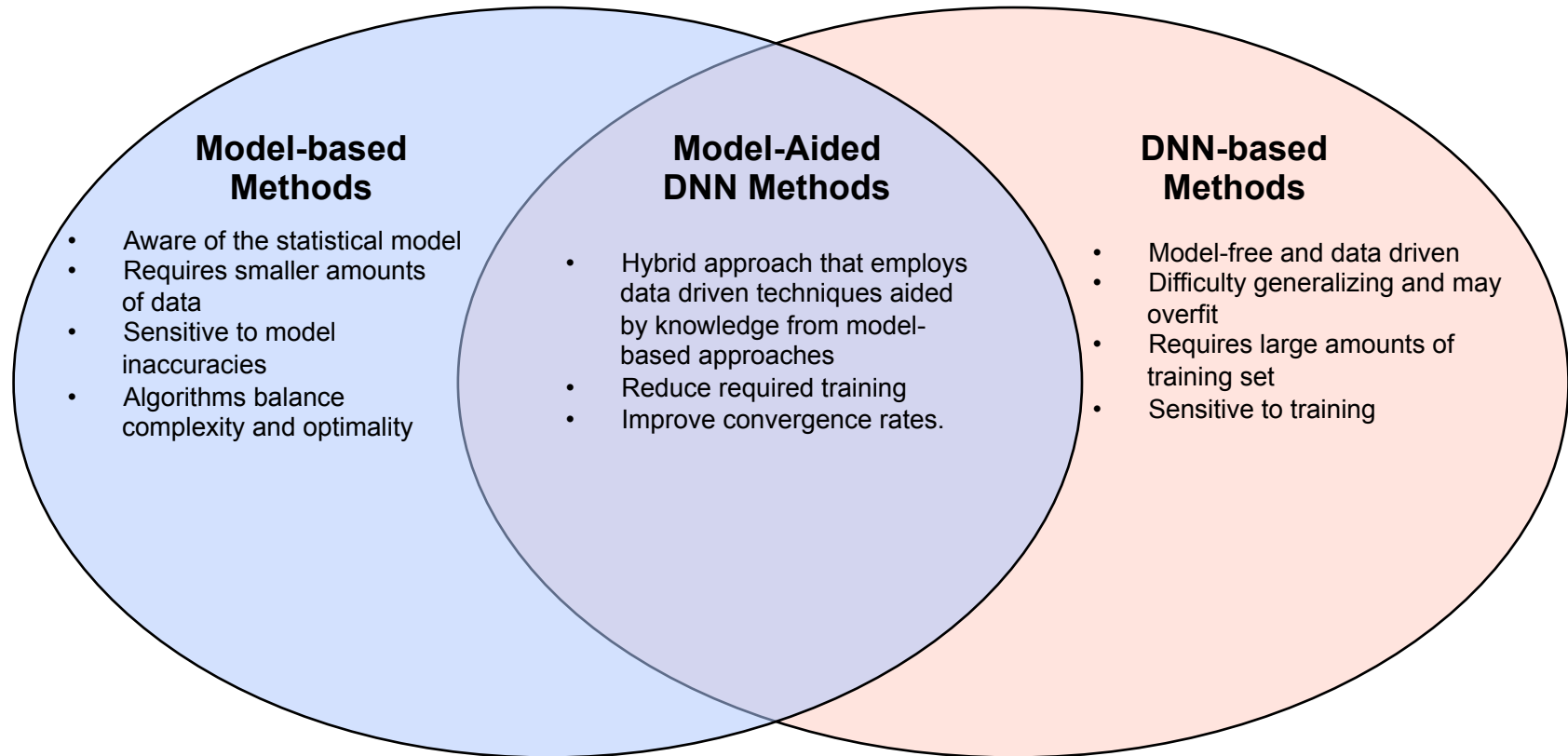
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Model Aided DNN Design

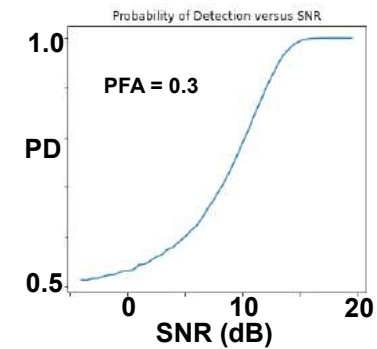
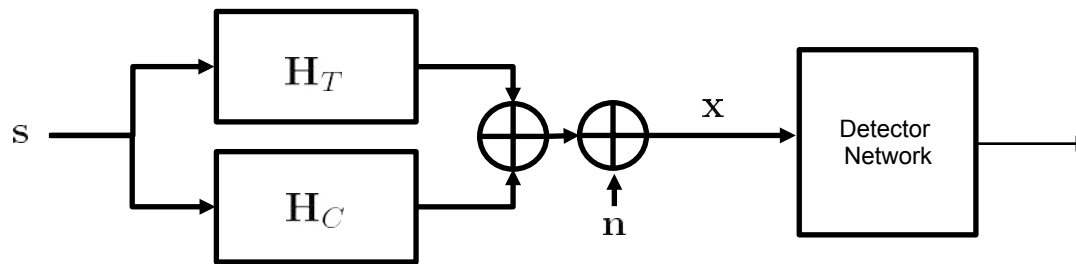




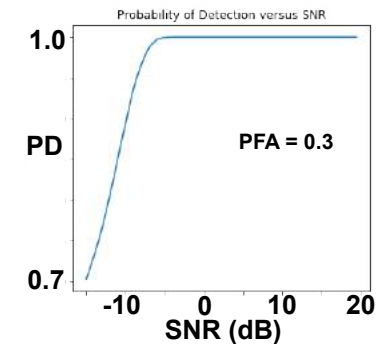
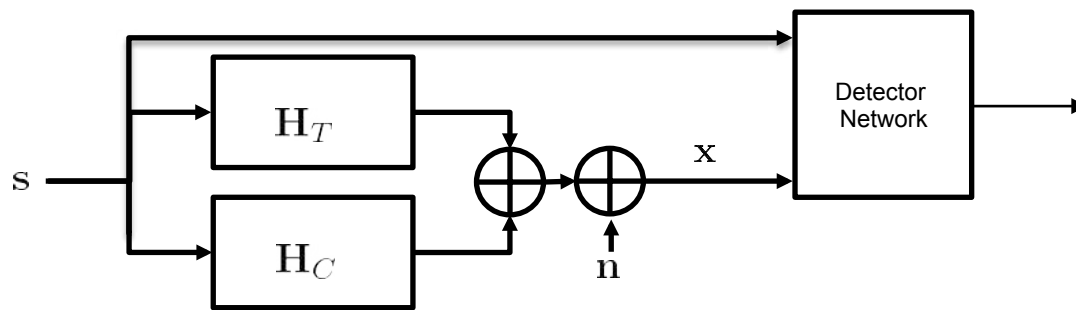
DNN Based Detector: First Attempts



- We tried training DNN to classify data as “target bearing” or “target free”:

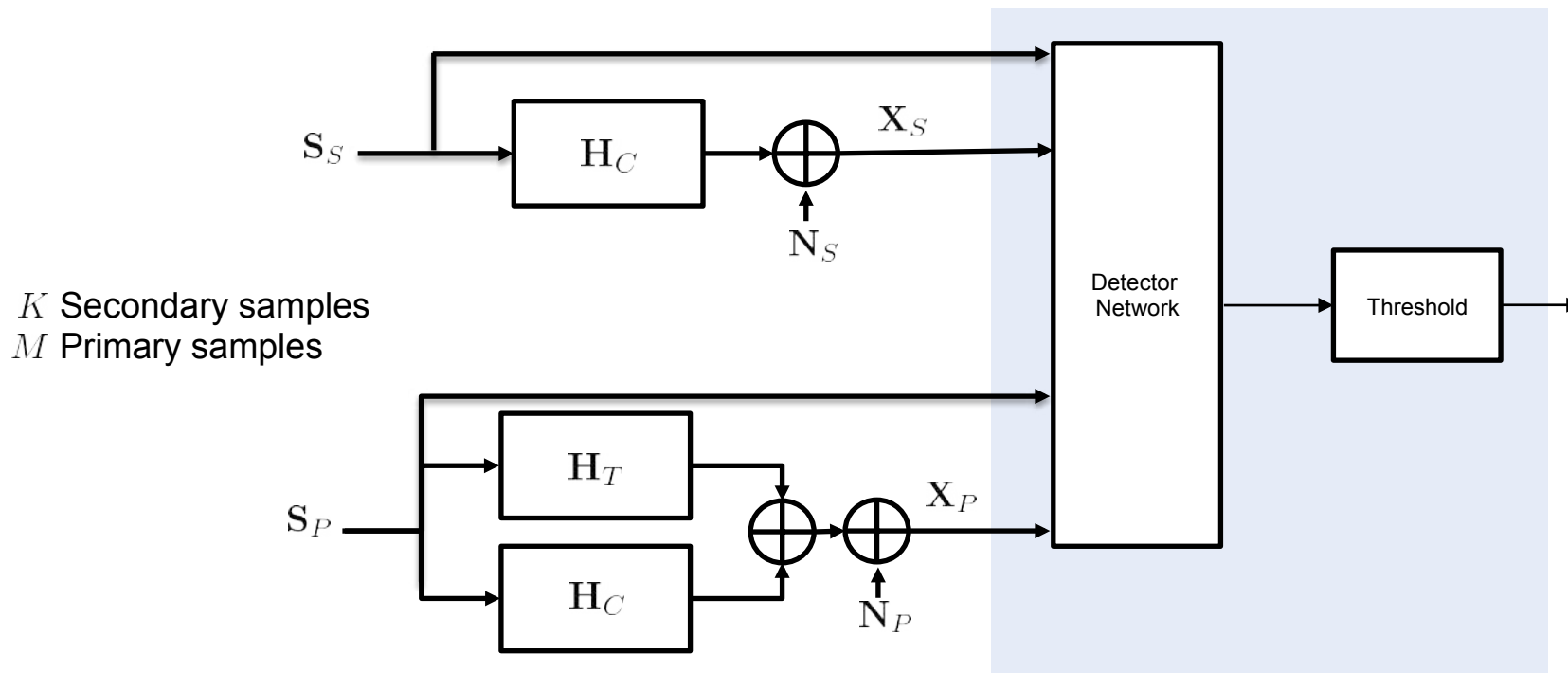


- GLRT exploits knowledge of waveform. Thus, we tried:





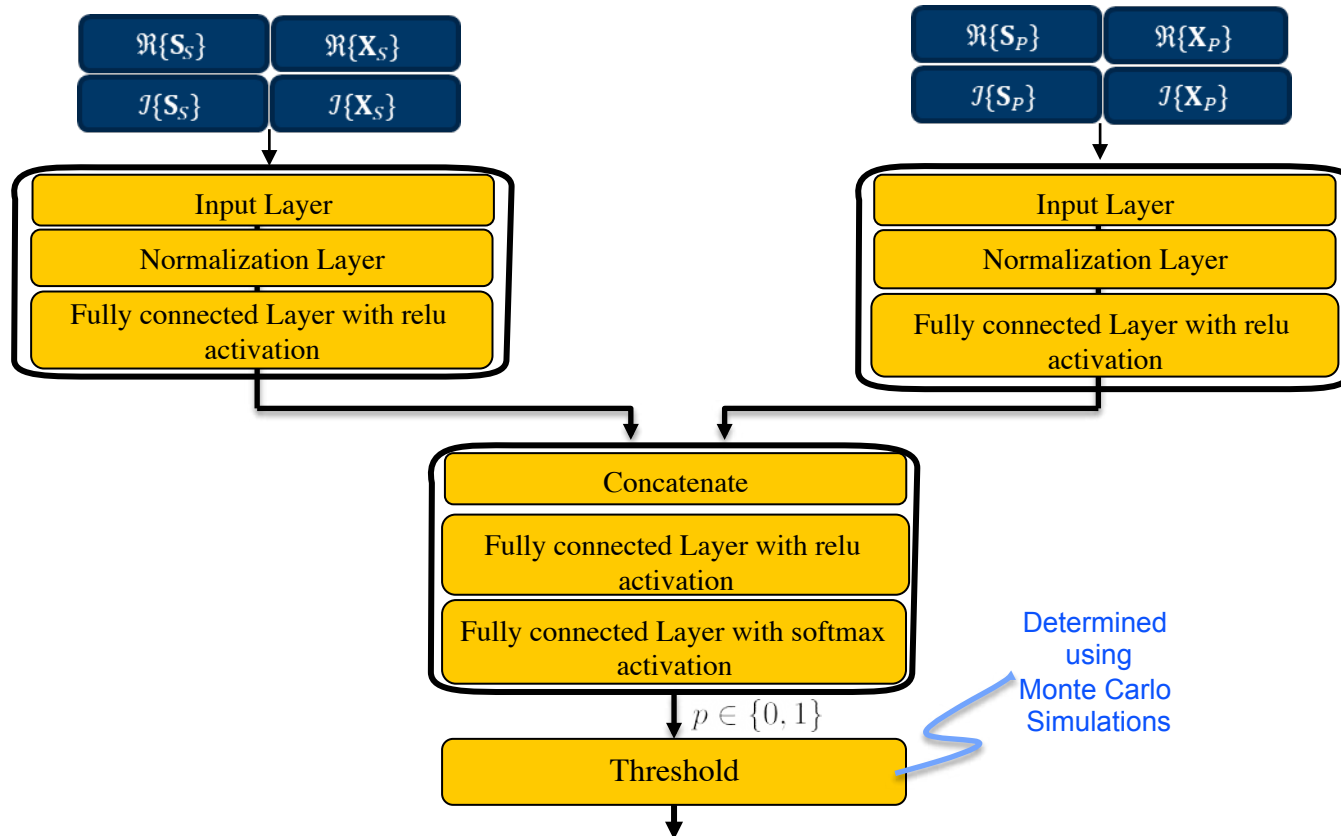
DNN Based Adaptive Detector: Architecture



- **Waveform history provided as input to the DNN**



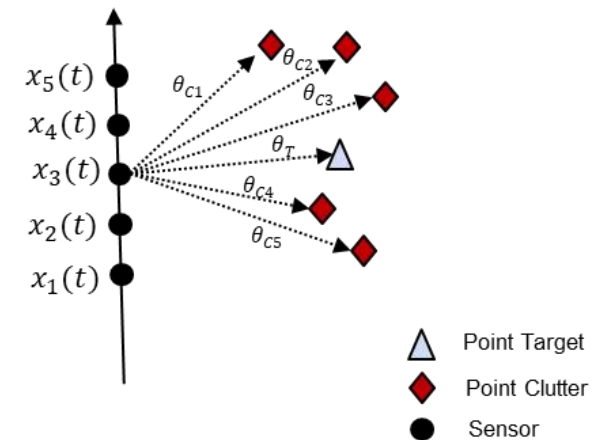
DNN Based Adaptive Detector: Network Layers





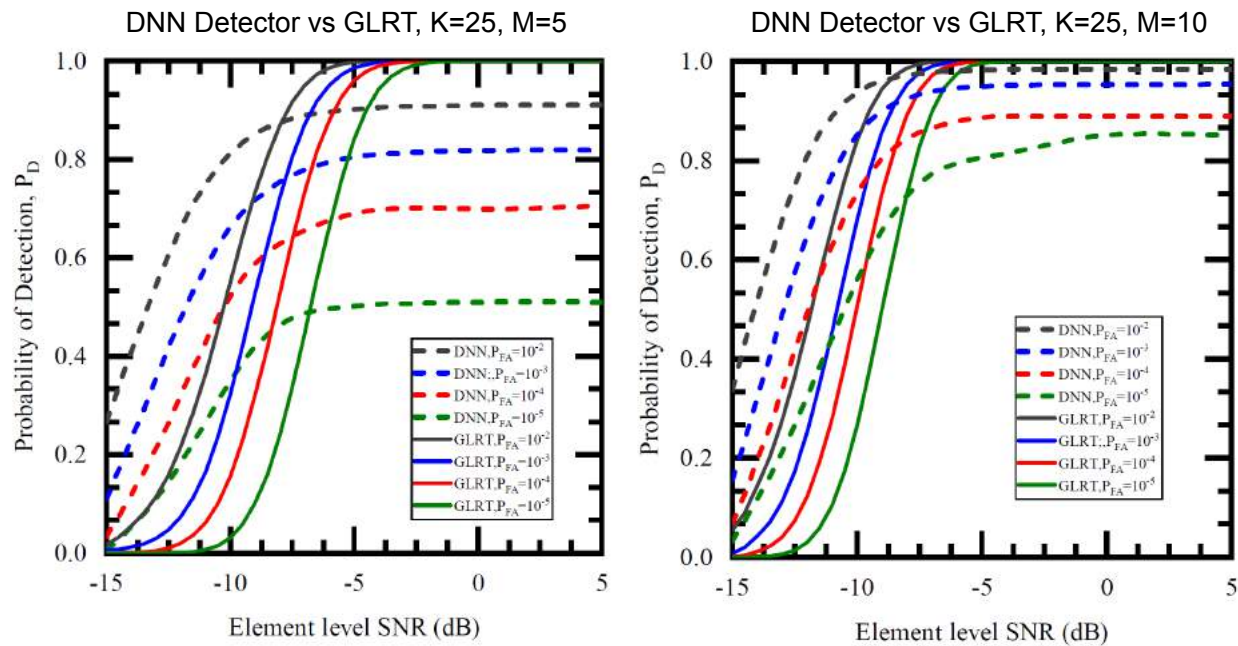
Simulation Results

- ULA of $l = 5$ sensors
- Single Point Target with zero doppler and five discrete point clutter placed randomly relative to the boresight of antenna array i.e. $\theta_T, \theta_{C_i} \sim \mathcal{U}(-\pi/4, \pi/4), i = 1, 2, \dots, 5$
- Zero mean Complex Gaussian Noise
- Rician Model of the Channel Matrices
- Waveform selected from Complex Gaussian distribution



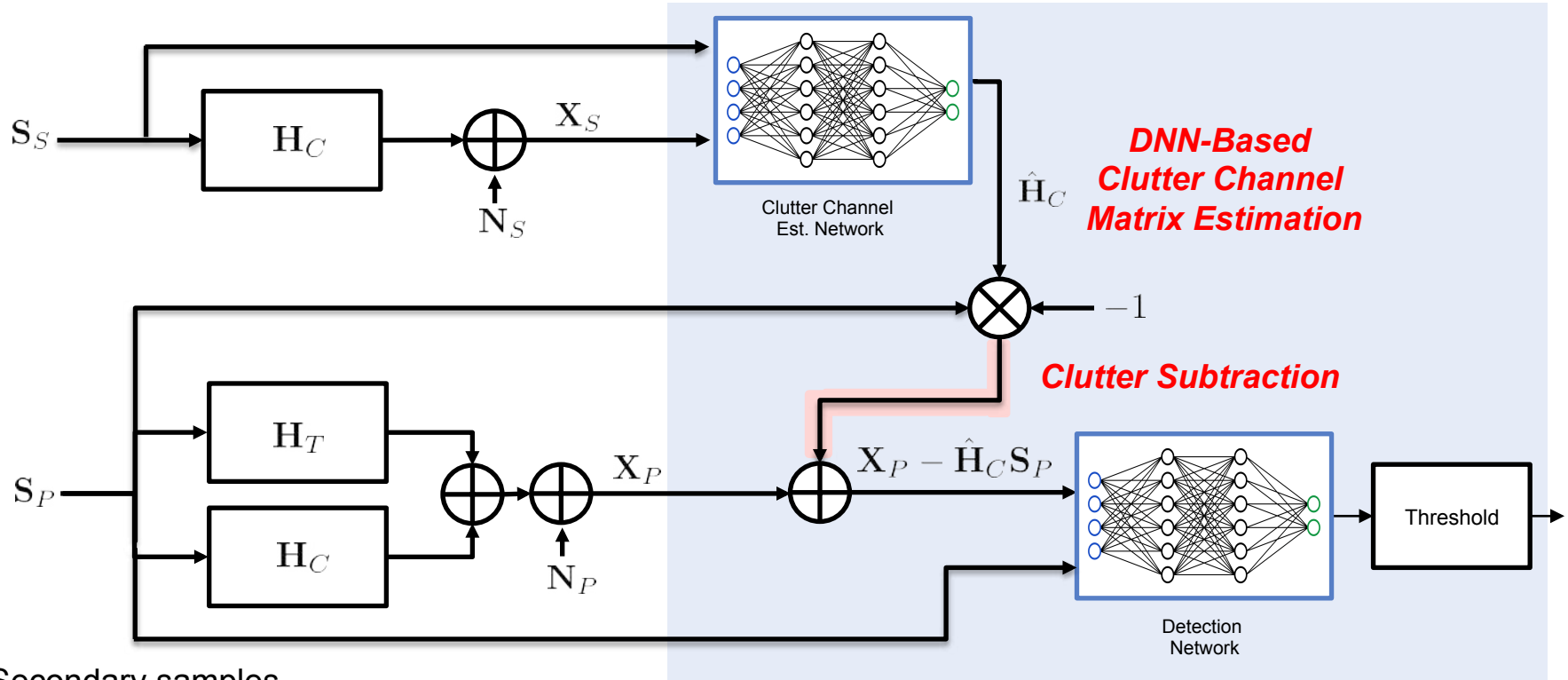


Simulation Results





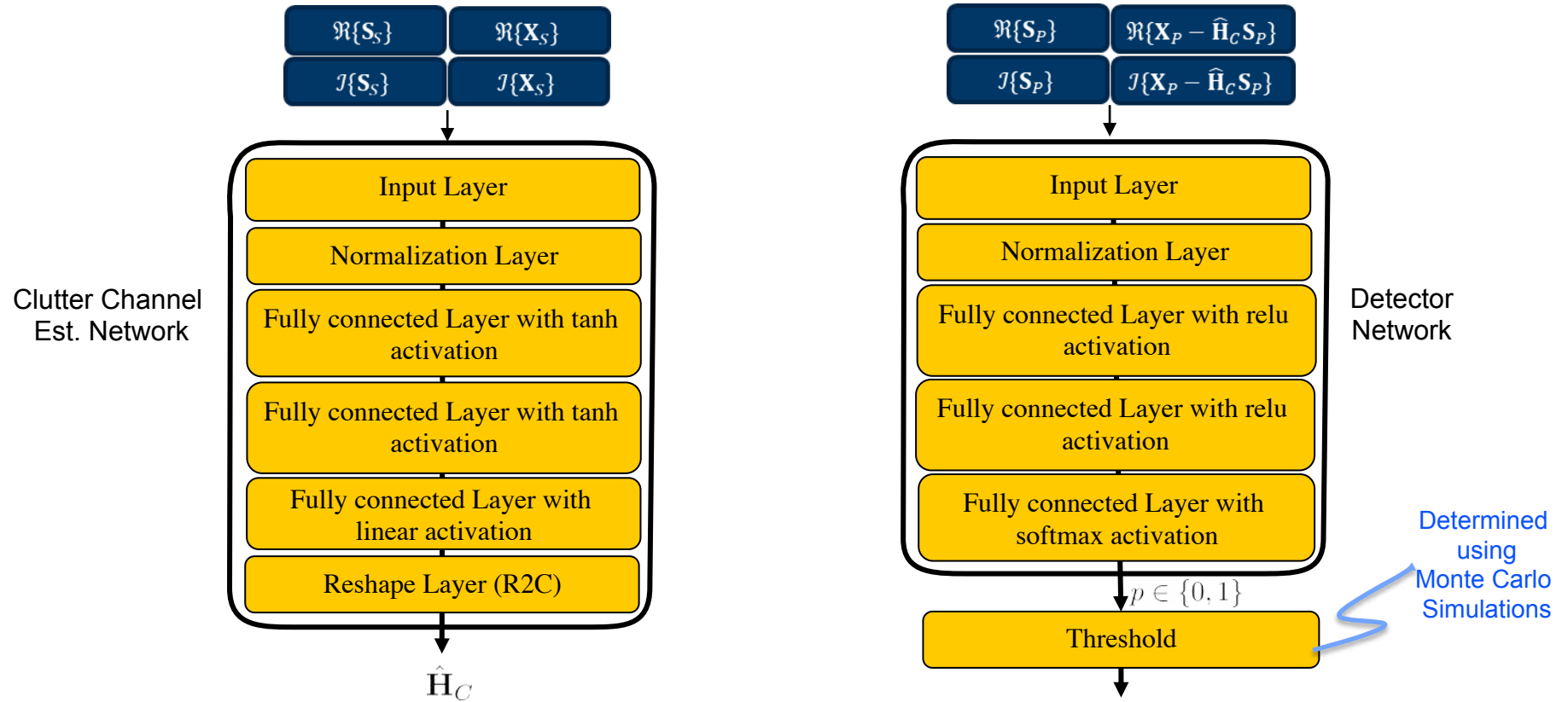
Model Aided DNN Detector: Clutter Removal



K Secondary samples
 M Primary samples



Model Aided DNN based Detector

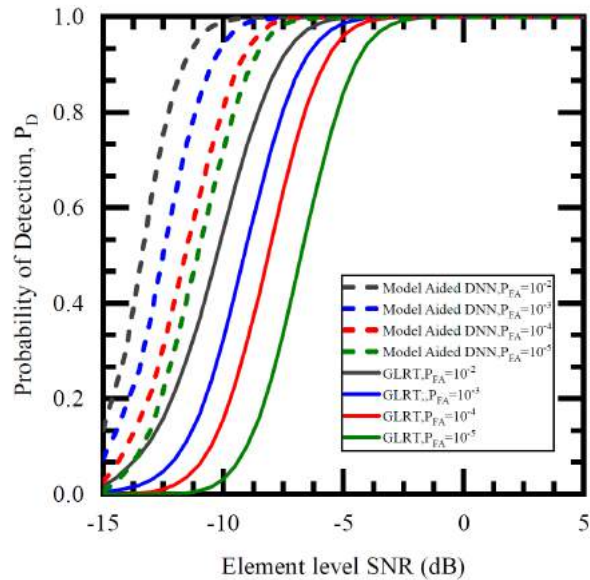




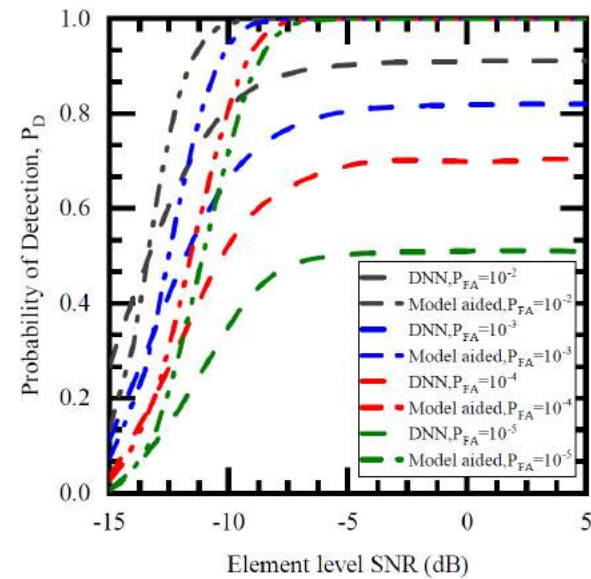
Simulation Results



Model aided DNN Detector vs GLRT,
K=25, M=5

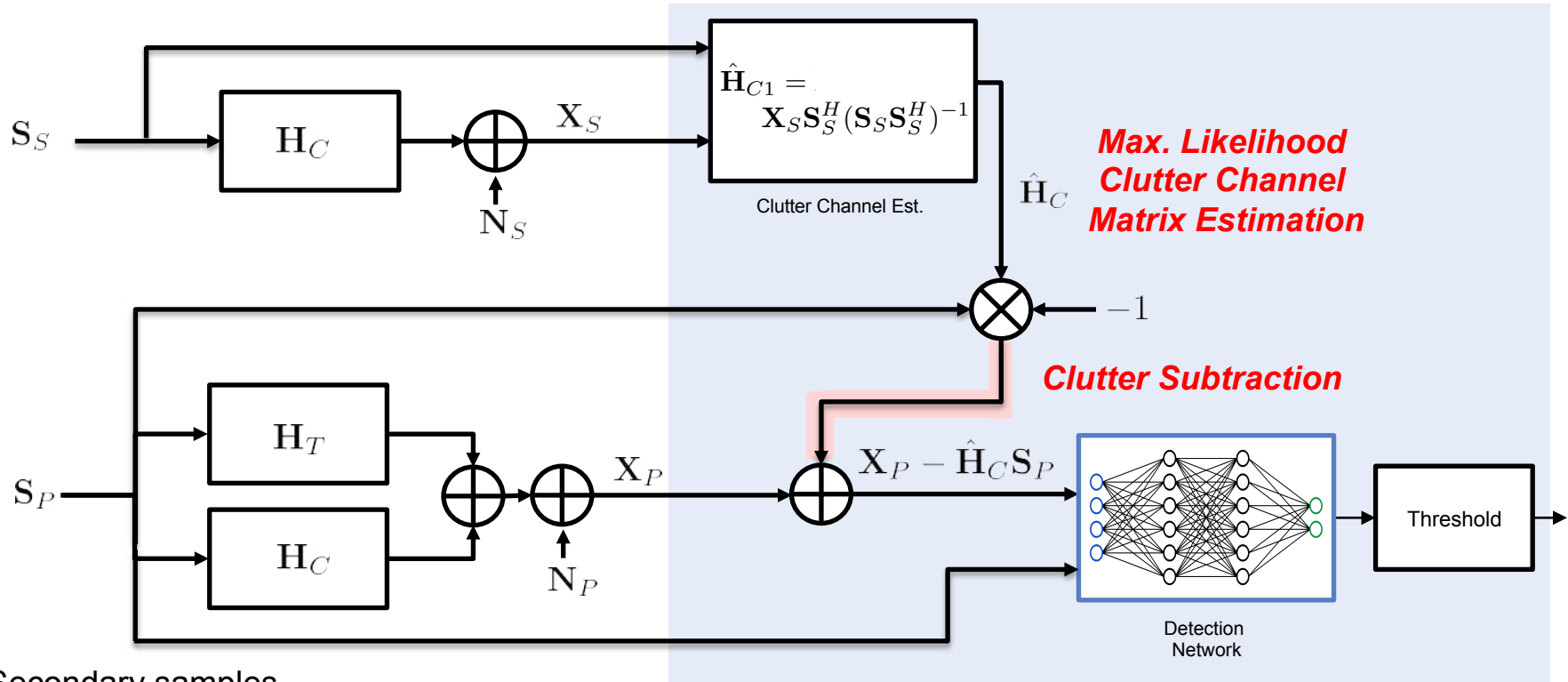


Model aided vs model free DNN
Detector, K=25, M=05





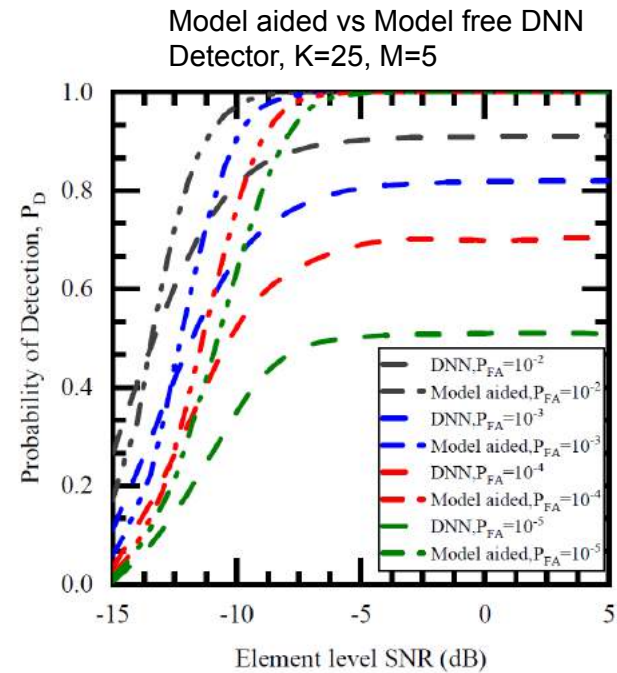
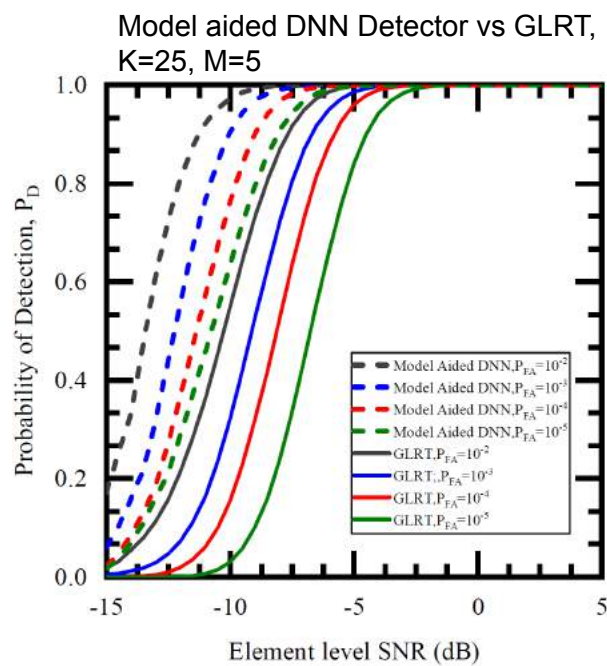
Model Aided DNN based Detector



K Secondary samples
 M Primary samples



Simulation Results





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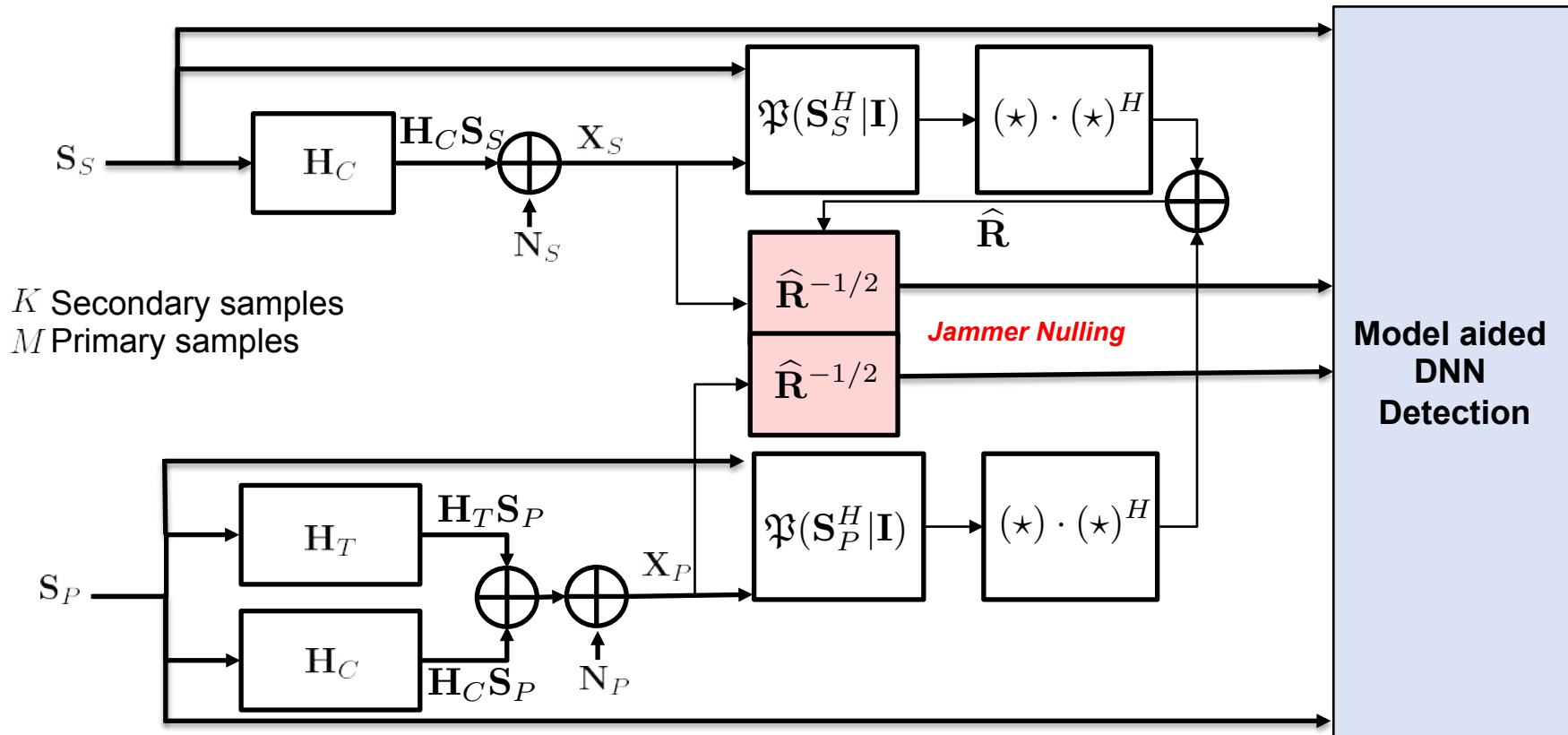
Summary



- **DNNs have difficulty generalizing without extensive amounts of training data**
- **Model-based GLRT makes efficient use of available data and performs robustly**
 - Uses knowledge of waveform
 - Performs clutter removal / nulling
 - Whitens / nulls waveform independent colored noise
- **DNN architecture modified to incorporate operations similar to GLRT results in significant improvement in DNN performance and convergence rate**
- **DNN can ultimately outperform GLRT after full “transfer learning”**



Cognitive Radar Detection: Next Step





Publications



- T. Ali and C. D. Richmond, "Optimal Target Detection for Random Channel Matrix-Based Cognitive Radar/Sonar," in 2021 IEEE Radar Conference (RadarConf21), 2021, pp. 1–6.
- T. Ali, A. S. Bondre, C. D. Richmond, "Adaptive Detection Algorithms for Channel Matrix-Based Cognitive Radar/Sonar," 2022 IEEE Radar Conference (RadarConf22), 2022, pp. 1-6.
- T. Ali, A. S. Bondre, and C. D. Richmond, "Model-Aided Deep Learning-Based Target Detection for Channel Matrix-Based Cognitive Radar/Sonar," The Journal of the Acoustical Society of America, Vol. 151, No. A100, ASA Meeting, May 2022.





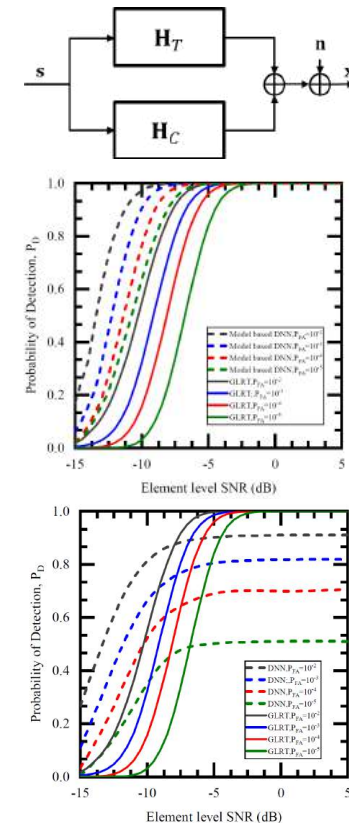
Questions & Comments Welcome



Summary



- **DNN-based target detection algorithm for channel matrix-based Cognitive Radar Framework**
- **Leveraging knowledge from GLRT derived for same framework**
- **Improvement in detection performance compared to model based GLRT algorithm and data driven DNN algorithm**
- **Future work:**
 - Making the architecture more robust to colored noise
 - Integrate DRL based transmitter





Adaptive Matched Filter (AMF) Detector



Form the optimal Neyman-Pearson test statistic, that is, the LRT.

Assume complex Gaussian statistics

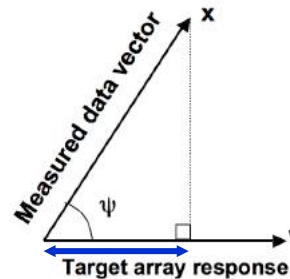
$$H_0 : p_{H_0} = \pi^{-N} |\mathbf{R}|^{-1} \exp[-\mathbf{x}^H \mathbf{R}^{-1} \mathbf{x}]$$

$$H_1 : p_{H_1} = \pi^{-N} |\mathbf{R}|^{-1} \exp[-(\mathbf{x} - S\mathbf{v})^H \mathbf{R}^{-1} (\mathbf{x} - S\mathbf{v})]$$

$$\longrightarrow \text{Likelihood Ratio Test} = \left[\frac{\max_S p_{H_1}}{p_{H_0}} \right] = t_{MF} = \frac{|\mathbf{v}^H \mathbf{R}^{-1} \mathbf{x}|^2}{\mathbf{v}^H \mathbf{R}^{-1} \mathbf{v}} \quad \begin{array}{l} \text{Matched} \\ \text{Filter} \\ \text{(Weiner Soln)} \end{array}$$

Since \mathbf{R} unknown use Sample Covariance:

$$\sum_{l=1}^L \mathbf{x}(l) \mathbf{x}^H(l) = \hat{\mathbf{R}} \longrightarrow \mathbf{R} \quad \longrightarrow \quad t_{AMF} = \frac{|\mathbf{v}^H \hat{\mathbf{R}}^{-1} \mathbf{x}|^2}{\mathbf{v}^H \hat{\mathbf{R}}^{-1} \mathbf{v}}$$



Known as the **Adaptive Matched Filter (AMF)** detector





Generalized Likelihood Ratio Test (GLRT)



Form the LRT based on the totality of data:

$$\begin{array}{l} \text{Test Cell} \\ \text{Training} \end{array} \quad [\mathbf{x} | \mathbf{x}(1) | \mathbf{x}(2) | \cdots | \mathbf{x}(L)] \triangleq \mathbf{X}_0$$

Assume homogeneous complex gaussian statistics

$$H_0 : p_{H_0} = \pi^{-N(L+1)} |\mathbf{R}|^{-(L+1)} \exp [-\text{tr} \mathbf{R}^{-1} \mathbf{X}_0 \mathbf{X}_0^H]$$

$$H_1 : p_{H_1} = \pi^{-N(L+1)} |\mathbf{R}|^{-(L+1)} \exp [-\text{tr} \mathbf{R}^{-1} (\mathbf{X}_0 - \mathbf{M})(\mathbf{X}_0 - \mathbf{M})^H]$$

$$\text{where } \mathbf{M} = [Sd | 0]$$

Maximize likelihood functions over all unknown parameters:

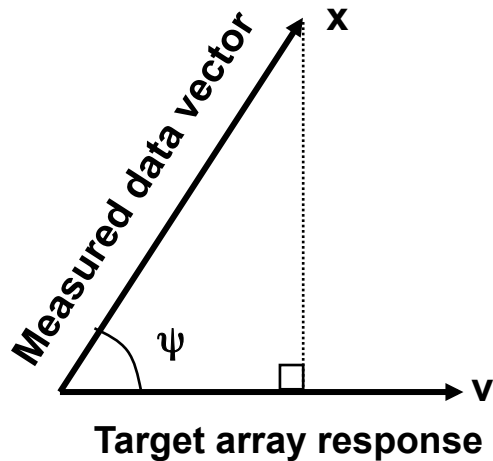
$$\rightarrow t_{GLRT} = \frac{\max_{S, \mathbf{R}} p_{H_1}}{\max_{\mathbf{R}} p_{H_0}} \Bigg|^{1/L+1} = \frac{1 + \mathbf{x}^H \hat{\mathbf{R}}^{-1} \mathbf{x}}{1 + \mathbf{x}^H \hat{\mathbf{R}}^{-1} \mathbf{x} - \frac{|\mathbf{v}^H \hat{\mathbf{R}}^{-1} \mathbf{x}|^2}{\mathbf{v}^H \hat{\mathbf{R}}^{-1} \mathbf{v}}}$$

Known as Kelly's / Khatri's GLRT

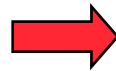




Adaptive Coherence Estimator (ACE)



- ACE statistic compares energy projected onto \mathbf{v} to total power in \mathbf{x}
- Inner product space defined wrt inverse of data covariance
 - in whitened space

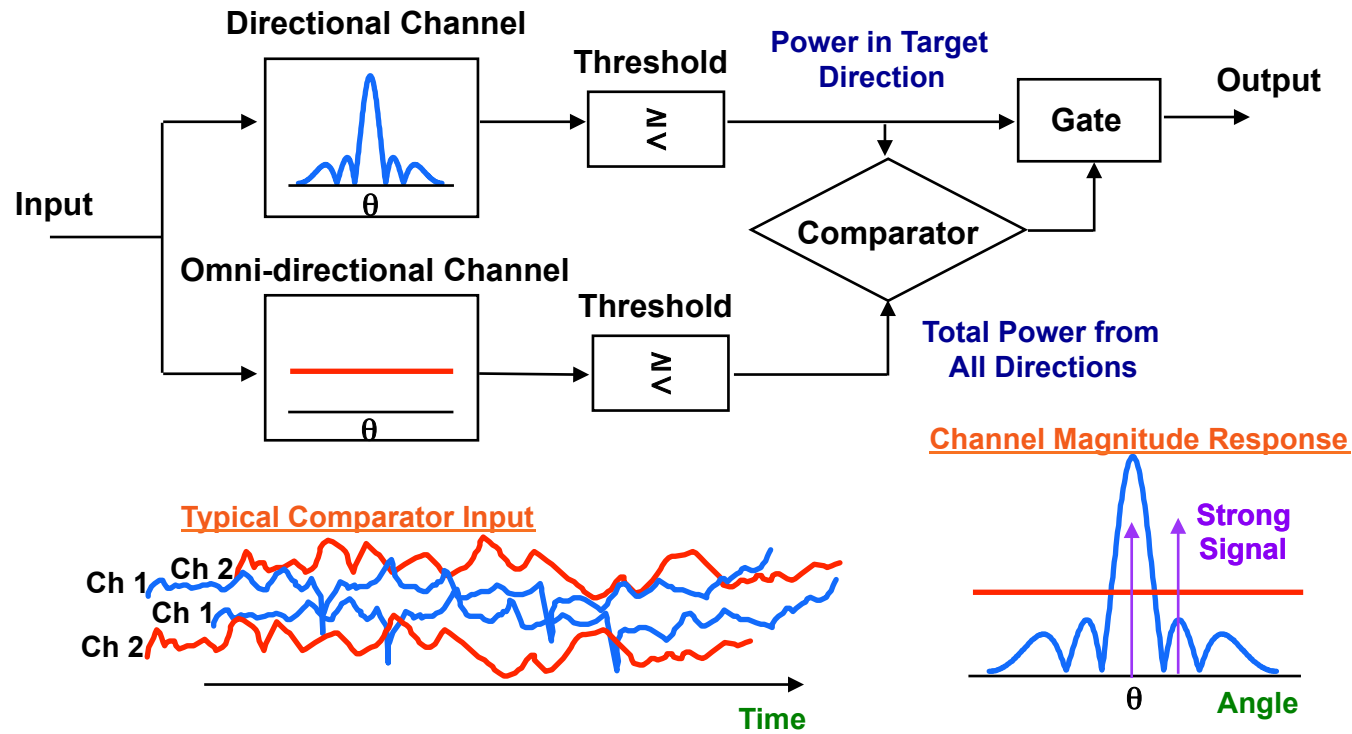


$$t_{ACE} = \frac{|\mathbf{v}^H \hat{\mathbf{R}}^{-1} \mathbf{x}|^2}{\mathbf{v}^H \hat{\mathbf{R}}^{-1} \mathbf{v} \cdot \mathbf{x}^H \hat{\mathbf{R}}^{-1} \mathbf{x}} = |\cos \psi|^2$$





Classical Sidelobe Blanking





2-D Adaptive Sidelobe Blanker (ASB) Detector



Step 1: Beamforming

$$t_{AMF} > \eta_{amf}$$

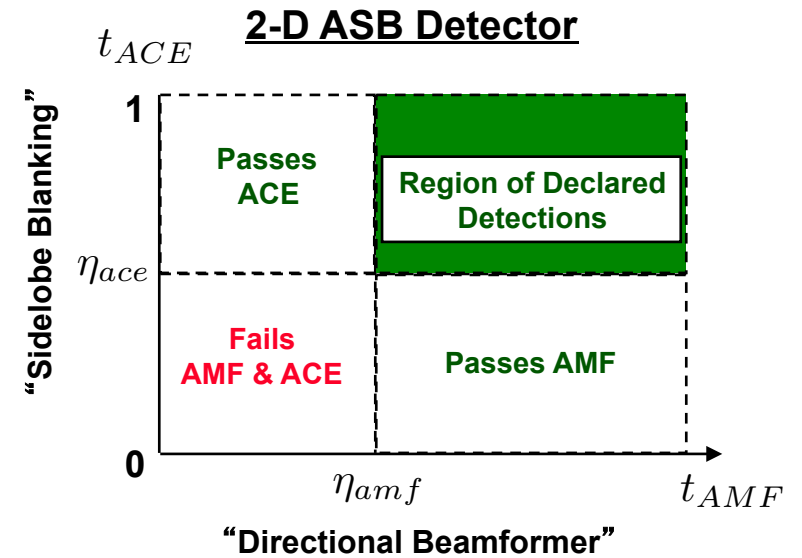
Power in Target Direction

Step 2 : “Sidelobe Blanking”

$$t_{AMF} > \eta_{ace} \cdot \mathbf{x}^H \hat{\mathbf{R}}^{-1} \mathbf{x}$$

Power in Target Direction

Total Power From All Directions





Model Free DNN based Detector



DNN Block	Layers
Branch 1 & 2	Layer 1: Input Layer with real and imaginary parts of the waveform and data vectors stacked Layer 2: Normalization Layer Layer 3: Dense layer with 128 perceptrons, activation: tanh Layer 4: Dense layer with 64 perceptrons, activation: tanh
Detection Block	Layer 1: Concatenation Layer Layer 2: Dense layer with 128 perceptrons, activation: relu Layer 3: Dense layer with 64 perceptrons, activation: relu Layer 4: Dense with 01 perceptron, activation: softmax
Threshold	Determined using Monte Carlo Simulations corresponding to particular PFA



Model Aided DNN based Detector



DNN Block	Layers
Clutter Channel Estimation Block	<p>Layer 1: Input Layer with real and imaginary parts of the secondary waveform and data vectors stacked</p> <p>Layer 2: Normalization Layer</p> <p>Layer 3: Dense layer with 128 perceptrons, activation: tanh</p> <p>Layer 4: Dense layer with 64 perceptrons, activation: tanh</p> <p>Layer 5: Dense layer with $2 \times l \times l$ perceptrons, activation: linear</p> <p>Layer 6: Reshape Layer</p>
Detection Block	<p>Layer 1: Input Layer with real and imaginary parts of the primary waveform and data vectors stacked</p> <p>Layer 2: Normalization Layer</p> <p>Layer 3: Dense layer with 128 perceptrons, activation: relu</p> <p>Layer 4: Dense layer with 64 perceptrons, activation: relu</p> <p>Layer 5: Dense with 01 perceptron, activation: softmax</p>
Threshold	Determined using Monte Carlo Simulations corresponding to particular PFA